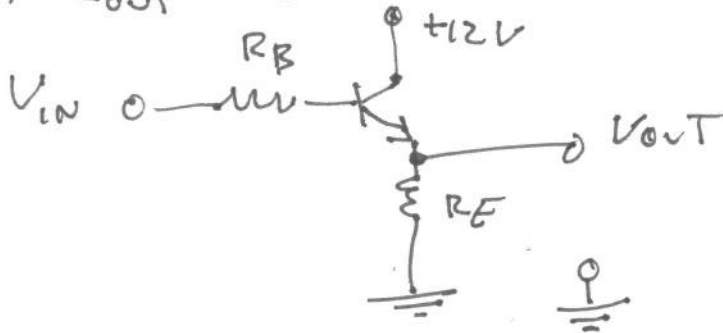


# INPUT / OUTPUT

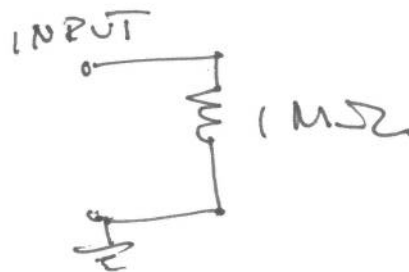
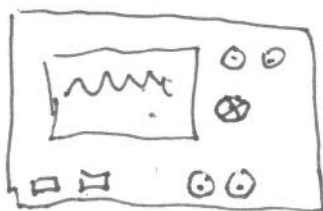
# IMPEDANCE

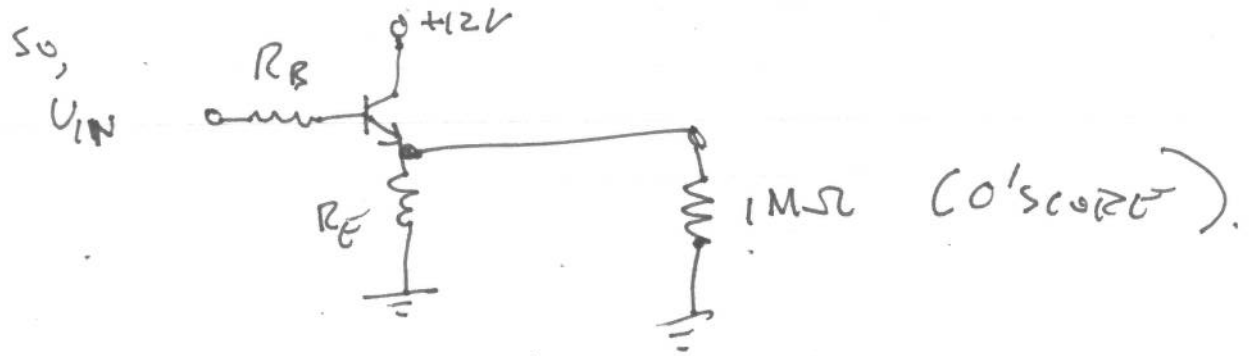
YOU TYPICALLY CONNECT CIRCUITS TOGETHER TO MAKE LARGER ONES. KNOWING THE INPUT/OUTPUT IMPEDANCE OF EACH SUB-CIRCUIT ALLOWS YOU TO UNDERSTAND BETTER THE BEHAVIOR OF THE LARGER CIRCUIT.

⟨⟩  $Z_{out}$  OF EMITTER FOLLOWER



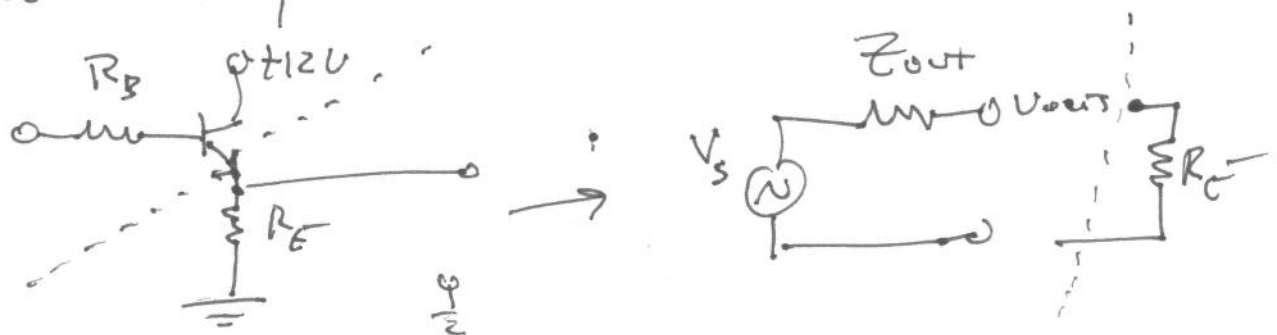
~~NOT~~ YOU MEASURE  $V_{out}$  w/ AN O'SCOPE. THE O'SCOPE CAN BE USEFULLY MODELED AS JUST A BIG ( $R = 1 M\Omega$ ) RESISTOR THAT DRAWS A GRAPH OF THE VOLTAGE DROP ACROSS THE  $1 M\Omega$  RESISTOR ON THE SCREEN.





SINCE  $R_E \parallel 1M\Omega$ , BUT TYPICALLY  $R_E \ll 1M\Omega$ , THE PARALLEL  $R$ 'S OF THE EMITTER FOLLOWER & THE SCOPE REDUCE TO JUST BEING  $R_E$ . THE EMITTER FOLLOWER DOESN'T NOTICE THE SCOPE'S PRESENCE. WE SAY THAT THE SCOPE "DOESN'T LOAD" THE CIRCUIT BEING TESTED.

THE EMITTER FOLLOWER CAN BE MODELLED MORE SIMPLY AS:



WHERE THE CIRCUITRY TO THE LEFT OF THE DASHED LINE HAS BEEN REPLACED W/ A VIRTUAL VOLTAGE SOURCE  $V_s$  & A VIRTUAL RESISTOR ("OUTPUT IMPEDANCE"). WE SAY "VIRTUAL" B/C THEY ARE NOT REALLY THERE, THEY JUST SEEM TO BE.

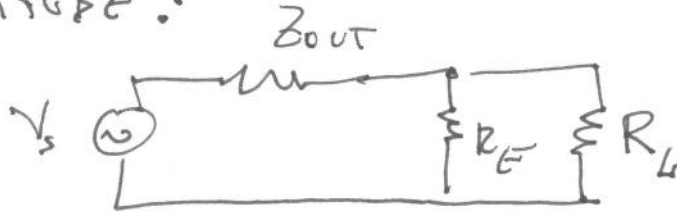
$Z_{OUT}$  IS A QUANTITY OF INTEREST. (2)

WE WANT TO MEASURE  $Z_{out}$ .  
 ( $V_s$  IS LESS INTERESTING.)

FROM PHYS 1304 LECTURE, THE VOLTAGE  
 DROP ACROSS  $R_E$  IS:

$$V_E = \left( \frac{R_E}{R_E + Z_{out}} \right) V_s$$

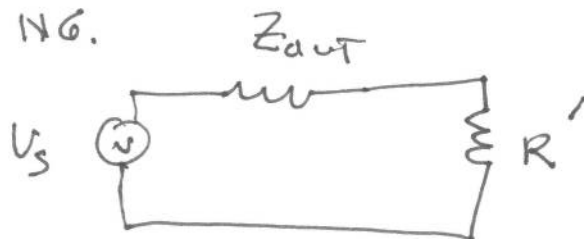
SINCE  $Z_{out}$  &  $V_s$  ARE UNKNOWN, WE  
 NEED ANOTHER EQUATION. ADD A  
 RESISTOR OF KNOWN VALUE IN PARALLEL  
 W/  $R_E$  AND OF ABOUT THE SAME  
 MAGNITUDE:



THE DROP ACROSS  $R_E \parallel R_L (= R')$  IS

$$V_{EL} = \left( \frac{R'}{R' + Z_{out}} \right) V_s \quad \text{w/ } R' = \frac{R_E R_L}{R_E + R_L}$$

REARRANGING.



DIVIDING OUR EQUATIONS FOR  $V_E$   
 &  $V_{EL}$  BY ONE ANOTHER ALLOWS US TO  
 SOLVE FOR  $Z_{out}$ .

$$\frac{V_E}{V_{EL}} = \gamma = \left( \frac{R_E}{R'} \right) \frac{R' + Z_0}{R_E + Z_0}$$

$$\gamma = \left( \frac{R_E}{R'} \right) \frac{R' + Z_0}{R_E + Z_0}$$

$$R' (R_E + Z_0) \gamma = R_E (R' + Z_0)$$

$$R' R_E \gamma + R' Z_0 \gamma - R_E Z_0 = R_E R'$$

$$Z_0 (R' \gamma - R_E) = R_E R' - R' R_E \gamma$$

$$Z_0 = \frac{R_E R' (1 - \gamma)}{R' \gamma - R_E}$$

SO, MEASURE  $V_E$  &  $V_{EL}$  TO GET  $\gamma$ .

$$\text{COMPUTE } R' = R_E \parallel R_L = \frac{R_E R_L}{R_E + R_L}$$

PICK  $R_L$ . I SUGGEST  $R_L = R_E$  BUT OTHER VALUES OK.

FOR THE EMITTER FOLLOWER, YOU WILL DISCOVER

$$Z_0 \approx R_B / \beta$$

WHERE  $\beta$  = TRANSISTOR  $\beta$ .

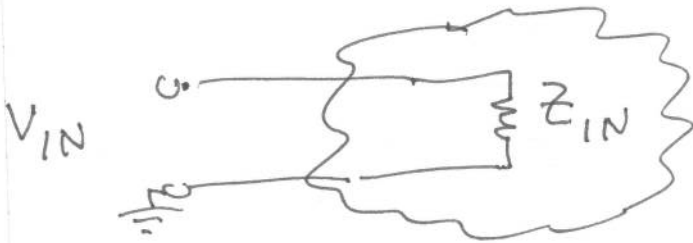
THIS TECHNIQUE OF MAKING  
2 MEASUREMENTS OF THE OUTPUT  
VOLTAGE, W/ 2 DIFFERENT RESISTOR  
VALUES CONNECTING THE CIRCUIT  
OUTPUT TO GROUND, IS COMPLETELY  
GENERAL. AND NOT RESTRICTED TO THE  
EMITTER FOLLOWER, YOU CAN FIND THE  
OUTPUT IMPEDANCE OF ANY CIRCUIT  
THIS WAY. (OR AT LEAST TRY. IT  
MAY BE TOO SMALL TO MEASURE. ▽)



THE CLOUD HIDES THE REAL CIRCUIT WHICH YOU MODEL AS JUST SOME VOLTAGE SOURCE IN SERIES W/ THE OUTPUT IMPEDANCE.

## INPUT IMPEDANCE

REPLACE ACTUAL CIRCUIT W/ CLOUD AND VIRTUAL RESISTOR (AKA, "INPUT IMPEDANCE")



TO MEASURE  $Z_{IN}$ , PLACE A RESISTOR OF YOUR CHOOSING (SAY,  $R = 1k\Omega$ ) IN SERIES WITH THE INPUT!

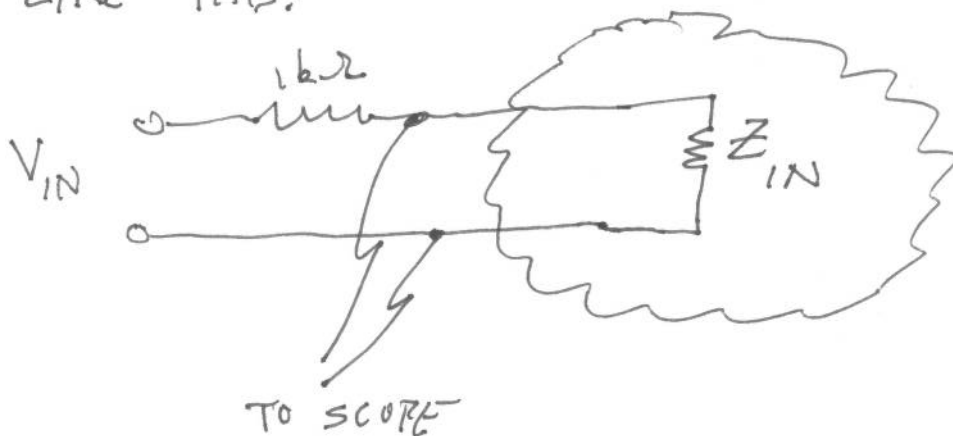


THE VOLTAGE DROP ACROSS THE  $Z_{IN}$  IS JUST

$$V_{Z_{IN}} = \left( \frac{Z_{IN}}{Z_{IN} + 1k\Omega} \right) V_{IN}$$

IF  $Z_{IN} \ll 1k\Omega$ , THEN  $V_{Z_{IN}} \approx 0$ .

FOR THOSE TOO EMBARRASSED TO ASK HOW YOU MEASURE THIS VOLTAGE DROP ACROSS  $Z_{IN}$ , CONNECT YOUR SCOPE PROBES LIKE THIS:



TO GET A PRECISE NUMBER FOR  $V_{Z_{IN}}$ , MEASURE  $V_{Z_{IN}}$ ,  $V_{IN}$  AND THEN

USE THE ABOVE EQUATION TO SOLVE FOR  $Z_{IN}$ . BTW, THERE IS NOTHING SACRED ABOUT THE  $1k\Omega$  RESISTOR. IT WAS SELECTED FOR CONVENIENCE. OTHER VALUES ARE OK.